# Investigation of Artificial Intelligence Algorithms for MPPT of Solar Photovoltaic System

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*Abstract:*- Solar energy has gained prominence as a primary renewable energy source for the generation of electricity in recent years. The maximization of power extraction from photovoltaic (PV) systems is a topic of significant interest due to the relatively low conversion efficiency of these systems. Therefore, a maximum power point tracking (MPPT) controller is essential in a PV system to achieve the desired output power. This paper implements three different MPPT controllers: sliding mode control (SMC), fuzzy logic controller (FLC), and artificial neural networks (ANNs). The performance of these controllers is evaluated on a PV system under varying irradiation and temperature conditions to analyze their ability to track the maximum power point (MPP). The results demonstrate that the SMC outperforms the FLC and ANN in terms of best performance with minimum oscillation under different operating conditions.

*Key-Words:*- Photovoltaic system, Boost converter, Sliding mode control, Artificial Neural Networks, Fuzzy Logic Controller, Simulation.

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# **1** Introduction

In today's world, there is a growing need, for energy sources, like power, wind energy, and biomass to meet the increasing electricity needs of consumers. These sources are pristine, abundant, and pollutionfree. Photovoltaic (PV) is a prominent clean energy source that efficiently powers electrical loads and may also feed energy back into the utility system. MPP operation is necessary because of the nonlinearity of the PV system's output. To ensure the continuous operation of MPP under varying ambient temperatures and solar radiation conditions, the use of control algorithms is necessary. It is difficult to determine which MPP has the smallest oscillation in proximity to the operational point, [1], [2], [3], [4].

Two ways to follow the MPP, the first is the conventional methods such as Fractional open circuit (OC) voltage, Fractional short circuit (SC) current, Incremental conductance (IC), and Perturbation and

observation (P&O). The main limitation of most of these earlier algorithms is their tendency to oscillate in the vicinity of the operating point in a steady state. Additionally, the direction of tracking the MPP is disrupted due to swiftly changing atmospheric conditions, [5], [6], [7], [8], [9].

The second is the intelligence methods such as FLC and ANN, SMC, and others. The ANN method must have undergone training using a significant number of solar irradiance and ambient temperature measurements. The FLC is sensitive to voltage variations and the current output from the PV panel. The system utilizes the derivative of power concerning current (dP/dI) and its derivatives as the inputs and calculates the duty cycle of the MPPT converter, [10], [11], [12], [13]. The SMC uses a step size that can be adjusted on the fly. When the operational point is significantly distant from the sliding surface, this method increases the step size to

expedite progress. Given that the effective point is near the sliding surface, the system will experience smaller step sizes, resulting in a shortened duration of oscillation around the maximum power point (MPP), [14], [15], [16], [17], [18]. The novelty can be summarized as:

- 1. Developing and adapting the FLC, ANN, and SMC based MPPT to optimize the power output of a PV with a resistive load.
- 2. Tracking of the MPP under different operating conditions during constant irradiance levels, rapid changes in irradiance levels, and rapid changes in temperature levels.

The organization of this paper will be as follows: Section 2 introduces the photovoltaic system. Section 3 describes the boost converter design. Sections 4, 5, and 6 present the MPPT controllers: SMC, FLC, and ANN, respectively. Section 7 shows the simulation results, while Section 8 presents the conclusion.

### 2 Photovoltaic System

The solar system under consideration, seen in Fig. 1, employs the PV system, whose electrical properties are listed in Table 1, a boost converter, a neural network based MPPT controller, and a resistive load, [19], [20], [21], [22], [23].



Fig. 1: Block diagram of the proposed system

Table 1. Parameters of PV	' module.
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Module Characteristics	Values
P <sub>max</sub>	220 W
V <sub>max</sub>	30 V
I <sub>max</sub>	7.35 A
I <sub>SC</sub>	8.81 A
V <sub>OC</sub>	36.8V

The single-diode variant strikes a good balance between ease of use and precision. In Figure 2, we see a photocurrent source connected in parallel with a nonlinear diode, a shunt resistor, and a series resistor. The photocurrent's origin is largely established by the cell's operational temperature and the amount of solar irradiation it receives. These equations characterize the PV cell model, [24], [25], [26]:

$$I = I_{ph} - I_D - I_{sh} \tag{1}$$

$$I = I_{ph} - I_S \left[ exp(\frac{q(V+IR_s)}{K_C A}) - 1 \right] - \frac{V+IR_s}{R_{sh}}$$
(2)

$$I_{ph} = [I_{SC} + K_1(T_C - T_r)]\lambda$$
(3)

Where  $I_{ph}$ : photocurrent current,  $I_D$ : diode current, I: output current from the cell,  $I_{Sh}$ : shunt resistor current,  $I_S$ : saturation current of the diode,  $K_1$ : Boltzmann constant, q: electron charge,  $T_c$ : actual temperature of the cell,  $R_{Sh}$ : shunt resistance, and  $I_{SC}$ : short-circuit current,  $T_r$ : reference temperature,  $\lambda$ : irradiance and  $R_{Sh}$ : series resistance, V and I are the output current and voltage of the PV module respectively.

#### **3** Boost Converter Design

The following clause describes the modeling of the dynamic behavior of the boost converter, [27], [28], [29], [30]:

When the switch is on:

$$V_{pv} = V_L = L \frac{di_L}{dt} \tag{4}$$

When the switch is off:

$$V_{pv} = V_L + V_o = L \frac{di_L}{dt} + V_o \tag{5}$$

The dynamic characteristics of the boost converter are from (6) and (7):

$$\frac{di_L}{dt} = \frac{V_{pv} - V_o(1-u)}{L} \tag{6}$$

Where  $V_{pv}$ : is the PV array voltage,  $i_L$ : the inductor current  $V_o$ : the converter output voltage, and  $V_L$ : is the converter inductor voltage.

## 4 Sliding Mode Control

SMC is a type of dynamic controller that is used to create resilient controllers for complicated, highorder, nonlinear dynamic plants that operate in uncertain settings. There are two fundamental concepts underlying SMC, the first creating a sliding surface that will function as the operational point in the initial stage. Secondly, within a finite amount of time, establishing a control law that shifts the effective point to a predetermined surface, [31], [32], [33], [34].

Begin by developing a surface within state space.

- The process of choosing a control law that compels the system's state path to approach a pre-established surface within a limited period.
- Sustaining it in proximity to this surface by employing suitable switching logic.

The SMC dynamic analysis: sliding surface proposition to be:

$$S = (V_{pv} - V_{ref}) \cdot G_1 + i_{Cin} \cdot G_2 \tag{7}$$

with G1 and G2 representing constant gains.

To implement the suggested method, (S) must equal zero and dS must be:

$$\frac{dS}{dt} = \left(\frac{dV_{pv}}{dt} - \frac{dV_{ref}}{dt}\right) \cdot G_1 + \frac{di_{Cin}}{dt} \cdot G_2$$
(8)

$$i_{Cin} = i_{pv} - i_L \tag{9}$$

$$i_{\rm Cin} = C_{in} \frac{dV_{\rm Cin}}{dt} \tag{10}$$

from (9) and (10) we obtain the following equation

$$\frac{dV_{pv}}{dt} = \frac{i_{pv} - i_L}{C_{in}} \tag{11}$$

Substituting (9) and (10) in (6)  

$$\frac{dS}{dt} = \left(\frac{i_{pv} - i_L}{c_{in}} - \frac{dV_{ref}}{dt}\right) \cdot G_1 + \left(\frac{di_{pv}}{dt} - \frac{di_L}{dt}\right) \cdot G_2 \quad (12)$$

Substituting (3) in (6)  

$$\frac{dS}{dt} = \left(\frac{i_{pv} - i_L}{c_{in}} - \frac{dV_{ref}}{dt}\right) \cdot G_1 + \left(\frac{di_{pv}}{dt} - \frac{V_{pv}}{L} + \frac{V_0(1-u)}{L}\right) \cdot G_2 = 0$$
(13)

Equation (7) thus characterizes the SMC dynamic model, in which the successful implementation of the

proposed method is ensured by the optimization of the fixed gains and parameters for the boost converter.

# 5 Fuzzy Logic Control

By regulating the duty cycle of the converter, the FLC targets to optimize the monitoring efficiency of the MPP of the PV panel. The inputs of the FLC correspond to the PV panel output current and voltage, whereas the output of the controller signifies the duty cycle. To optimize power output, the FLC controls the voltage through the converter duty ratio, which is determined by the power ratio (dp/dv) change. The FLC is utilized to regulate the DC converter and accepts duty cycle (CD), error (E), and change in error (CE) signals as inputs. The parameters for the input and output are as follows, [35], [36], [37]:

$$E(k) = \frac{P(k) - P(k-1)}{V(k) - V(k-1)}$$
(14)

$$CE(k) = E(k) - E(k-1)$$
 (15)

$$CD = f(E, CE) \tag{16}$$

Where the immediate power (P(k)) and instant voltage (V(k)) of the PV are denoted as such.

The direction in which the load target point at instant k is moving is denoted by the change in error, while the input error E(k) shows whether the target point of the load is to the right or left of the MPP of the PV panel. The controller's design comprises three fundamental components: defuzzification, base rule, and fuzzification, [38], [39], [40], [41], [42]. The input and output membership functions of the controller are illustrated in Figure 2. The FLC rule includes 25 rules, as described in Table 2.



Fig. 2: Fuzzy system membership function. (a) E. (b) CE. (c) CD

E	CE				
	NB	NS	ZE	PS	PB
NB	PS	PB	NB	NB	NS
NS	PS	PS	NS	NS	NS
ZE	ZE	ZE	ZE	ZE	ZE
PS	NS	NS	PS	PS	PS
PB	NS	NB	PB	PB	PS

Table 2. Rules of the FLC

## 6 Artificial Neural Network

There are several ways in which AI methods excel over more conventional methods. Traditional systems have limitations including being slow to adapt to shifting solar temperature and irradiance, and occasionally failing to keep tabs on the peak power point. Input data (such as irradiance and temperature) is received by the input layer, and from there is sent to the second and third layers via a network of hidden neurons. The converter's output (duty cycle) is supplied by the third layer. To determine the MPP, a three-layer neural network is illustrated in Figure 3. The input layer is responsible for receiving input data, which consists of temperature and irradiation. Several concealed neurons are located in hidden layers, from which the input layer transmits data to the third layer. The duty ratio output is supplied to the converter by the third layer. The inputs and outputs are, [43], [44], [45], [46], [47]:

$$Y_j^h = f\left(\sum_{i=1}^N W_{ij}X_i + \theta_j^h\right) \tag{17}$$

$$Y_k^o = f\left(\sum_{j=1}^N W_{kj} Y_j^h + \theta_k^o\right) \tag{18}$$

Where  $W_{ij}$  and  $W_{kj}$ : The lines linking the three levels are allocated weights;  $\theta_k^o$  and  $\theta_j^h$ : The bias values associated with the concealed layer and output, respectively;  $X_i$  and  $Y_j^h$ : The signal values about the input and output lines.



Fig. 3: The construction of the ANN

## 7 Results and Discussion

The simulations utilized in this research were carried out through the implementation of systems in MATLAB, which execute the suggested SMC, ANN, and FLC. The output PV power of the SMC, FLC, and ANN using constant conditions (T = 25C, G =  $1000W/m^2$ ) is depicted in Figure 4. The SMC exhibits superior characteristics in terms of rise time, response time, and absence of overshoot when compared to the ANN and FLC, [48], [49], [50], [51], [52], [53].



Fig. 4: An analysis of the three controllers in comparison to a constant atmospheric condition

For the three controllers, Figure 5 depicts the PV output power when the irradiance drops abruptly from 1000 W/m<sup>2</sup> to 800 W/m<sup>2</sup>, then to 600 W/m<sup>2</sup>, and ultimately to 400 W/m<sup>2</sup>. They possess an exceptional capacity to monitor the MPP by the four controllers. However, the rise and settling periods of the SMC are shorter.



Fig. 5: The power output of the three controllers at a constant 25 C and varying levels of irradiance.

The PV output power for all three controllers is depicted in Figure 6, with temperature fluctuations and irradiance of  $1000 \text{ W/m}^2$  held constant. The temperature fluctuated between 25C and 30C, then 35C, 40C, and 25C once more. It is noteworthy that they possess an exceptional capability to monitor the MPP for every controller. However, the rise and settling periods of the SMC are shorter.



Fig. 6: The power output of the three controllers as the temperature and irradiance levels remain constant.

#### 8 Conclusions

This paper describes various MPPT control algorithms utilized to determine the MPP of a PV panel subjected to variable conditions. The SMC, FLC, and ANN were devised to monitor the MPP across diverse atmospheric conditions. To assess the effectiveness of the controllers being evaluated, a variety of scenarios were investigated. These scenarios comprised consistent irradiance levels, precipitous fluctuations in irradiance levels, and abrupt temperature changes. In contrast to FLC and SMC, the results demonstrate that the SMC can monitor the reference rapidly and with superior performance. At a steady condition, the MPP is devoid of any oscillation.

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