

# Quadrature Sandwich Rectenna for Wireless Power Transfer

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*Abstract:* This novel work presents a quadrature sandwich rectenna circuit for collecting RF from four input ports at different angles. The scheme's high efficiency and increase in output power are significant for the operation of wireless electronic devices. This work demonstrates the requirements of self-powering. Specifically, simulation efficiency is shown to achieve 77%, and input power is 0 dBm for a 5 K $\Omega$  resistance load and dual-band operation. The results of this experiment are encouraging in terms of wireless powering of some devices in wireless telecommunication systems. This scheme can also be applied to radio frequency wireless power transfer and harvesting, and would be useful in industrial, scientific, and medical applications. Furthermore, its mode of operation is sustainable, and it is compact.

*Key-words:* Wireless power transfer, energy harvesting, power combiner, printed circuits, rectifiers.

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## 1. Introduction

Base transceiver stations (BTSs) and cellular towers are widely used to feed wireless devices and are parts of advanced telecommunication systems. Their function is to send and receive waves via radio frequency (RF) and microwave technology, as described in [1, 2]. RF and microwave radiation are vital for transmitting information and energy to wireless devices, with BTSs keeping these devices continuously active.

Wireless receiver circuits have become widespread due to technological developments. These receiver circuits require continuous supplies of energy for sustained operation. Therefore, researchers have been working to produce solutions for the self-powering of wireless devices from RF signals. One of these solutions is energy harvesting (EH), which is defined as a process of reaping alternating current (AC) from external sources such as RF and converting it to direct current (DC).

Due to effective radiated power (ERP), approximately 50 dBm per channel can be transmitted, as stated in [1, 2]. This makes the conversion of RF to EH attractive due to its availability and cost-effectiveness. These aspects mean that it is practical for receiver circuit applications.

The study in [3] presented a receiver circuit that integrated an antenna and rectifier using a single diode. Its frequency of operation was from 2 to 18 GHz at an input power of between -17 and 15

dBm. It obtained an efficiency of 20% with a resistance load of 100  $\Omega$  at 3 GHz.

The experimental work in [4] designed a rectenna consisting of a slot-loaded dual-band folded dipole antenna and a dual-band rectifier. The resistance load was 2.2 k $\Omega$  and the input power was -9 dBm. The authors [4] achieved measurement efficiencies of 37% and 30% at 915 MHz and 2.45 GHz, respectively.

In [5], a dual-frequency scheme was proposed with ultralow-power, using a single diode rectifier topology. The input power levels were -11 and -13.5 dBm at 915 and 2450 MHz, respectively. The output DC voltage was 200 and 313.5 mV, respectively, with a resistance load of 2200  $\Omega$ .

Another article presented a new rectenna topology to improve power handling [6]. It increased the number of rectifying branches with a single collected DC output, and the topology was tested using a two-branch rectifier. The power conversion efficiency was 57% at 17 dBm, for an operating frequency of 2.1 GHz.

The work in [7] demonstrated the concept of dual-band resistance compression networks (RCNs) for rectifier circuits with enhanced performance. Its proposed rectifier circuits changed the input power level and variations in the rectifier load. Further, the frequencies were a dual-band 915 MHz and 2.45 GHz rectifier based on an RCN. This reduced the sensitivity to variations in the output load and input power.

In [8], a circuit receiver was designed for self-power harvesting and wireless connection to an ambient base station. It attained efficiency of 54.81% at 0 dBm input power. In other work, the researchers in [9] considered the time domain superposition coding technique, adding on the G3-PLC standard and performing reproduction tests. Their outcomes show that the framework had a gain of up to 4 dB when the G3-PLC standard was added with time area superposition coding. However, the calculations could be challenged on the obstruction of high commotion and the adequacy of multipath impact.

A method was presented in [10] to reuse current for a wideband low noise amplifier (LNA) and ultra-wide band (UWB) purposes. It was completed by the wideband input matching and represents a collection of degenerative parallel LC circuit and resistive returns in a shunt-shunt connection. Additionally, the cascaded LC method was applied in [10] to obtain results matching.

The above-mentioned state-of-the-art approaches addressed some issues of converting RF to DC for EH. However, there are several challenges involved in increasing DC output power. These problems are partly associated with combining RF with the rectifier's output power. Thus, the input power and the distance between the transmitter and receiver remain unresolved issues that affect the achieving of high power conversion efficiency (PCE).

In addressing this research gap, the present paper proposes a prototype to obtain proper results. The prototype achieves more bandwidth, higher efficiency and greater compactness, and is robustly packed in a box shape to protect its components from the elements.

The circuit in the proposed prototype consists of two layers in the form of a parallel T-junction power combiner coupler for combining RF. Each T-junction power combiner coupler is connected to an L-type matching network. A pair rectifier voltage doubler is then used to combine separate DCs into a single output port. It is expressed as a quadrature input antenna RF energy reception angle side, with dual DC power. This is a multi-input, single-output (MISO) circuit that is shaped so that the elements are inside the box. The purpose is to combine RF reception, protect the elements, and combine DC outputs, as shown in Fig. 1.

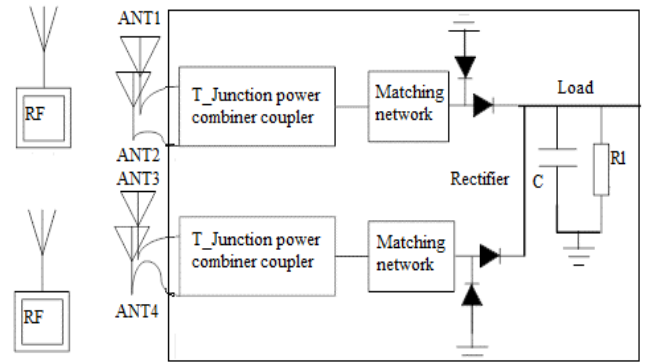


Fig. 1: System architecture, consisting of BTSs and the proposed circuit.

## 2. Circuit Design and Simulation

For the design, RO4350B laminate from Rogers Company was chosen, due to its low loss, availability, and low cost. Table 1 shows the parameters of the substrate. Several variables are used in the design of the circuit, including speed of light  $c = 3 \cdot 10^8$  m/s, frequency  $f = 5$  GHz, and wavelength  $\lambda = 0.06$  m. Advanced Design System (ADS) software was used to design the circuit, and harmonic balance was used to simulate the analogue RF and microwave circuit for the prototype.

The parameters of the T-junction power combiner coupler are shown in Table 2. This scheme can be implemented using a transmission line (TL).

Table 1. Parameters of Substrate RO4350B [8]

Relative permittivity ( $\epsilon_r$ )	Thickness ( $h$ )	Thickness of copper line ( $T$ )	Loss tangent ( $TanD$ )
3.48	1.52 mm	17.5 $\mu$ m	0.004

Table 2. Component Parameters

Characteristic impedance $Z_0$	50 $\Omega$
$Z_0 \cdot \sqrt{2}$	70.7 $\Omega$
TL width (W) of two input ports	2 mm
TL length (L) of two input ports	3 mm
TL $W = L$ of one output port	3 mm
Angle of each curve of TL	90°
Curve W	1.89 mm
Curve L	2 mm
TL to connect each curve with input port or output port $W = L$	2 mm

Table 3 shows the parameters for a SPICE model pairs diode from a Skyworks Solutions datasheet.

Table 3. SPICE Model of SMS7630-006LF [11]

Parameter	Unit	Value
Is	A	5E-6
Rs	$\Omega$	20
N	-	1.05
TT	Sec	1E-11
CJO	pF	0.14
M	-	0.40
EG	eV	0.69
XTI	-	2
FC	-	0.5
BV	V	2
IBV	A	1E-4
VJ	V	0.34

Figure 2 shows a schematic of the proposed system that consists of two layers of a T-junction power combiner coupler in parallel. Each T-junction is connected with a matching network and then connected with a voltage doubler integrated with the load stage.

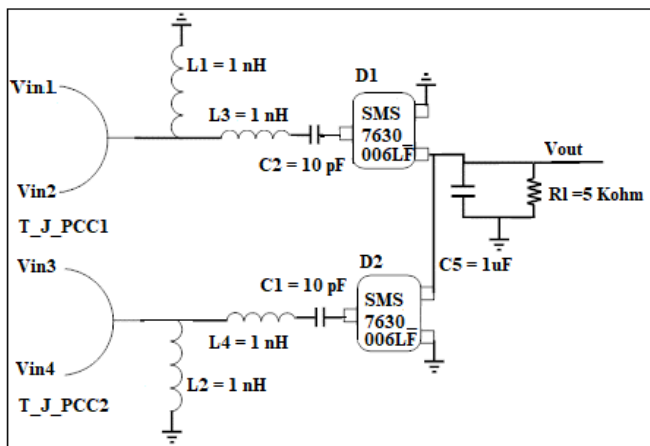


Fig. 2: Schematic of a two-layer T-junction power combiner coupler integrated dual voltage doubler.

Figure 3 shows the schematic of Schottky diode model SMS7630-006LF for the voltage doubler with the matching network. The output of the rectifier stage was combined with a capacitor to eliminate ripples in the load stage. This effectively increases the DC output power.

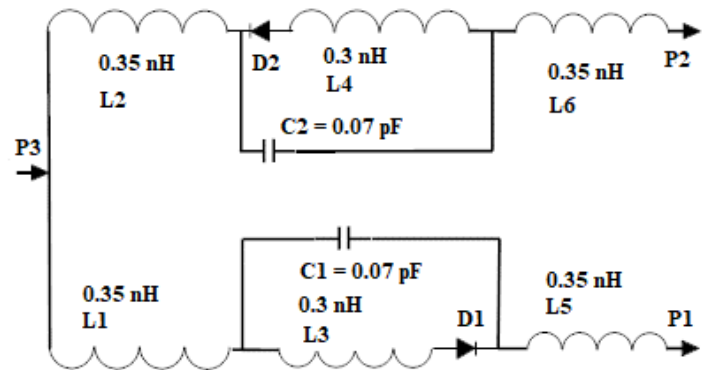


Fig. 3: Schematic of diode SMS7630-006LF.

Figure 4 plots the simulated DC output power (dBm), as seen in Fig. 2. Four generators are connected to the receiver circuit in order to combine the DC outputs via a rectifier. A 0 Hz component shows output power of  $-1.7$  dBm, while all four generators combined show input power of 0 dBm, or 0.25 mW for each generator. The proposed structure is designed to reduce losses and has more electromagnetic waves captured by a two-dimension dual antenna. Using this design, the losses were reduced by 50% compared with our previous work [8]. Even when using an array of four antennas and a single detector, the simulation figured out more losses, with output power at  $-6.6$  dBm. These results were calculated using Equation (2.1), which expresses the output power as follows:

$$\text{Output power (dBm)} = 10 \log_{10} (0.5 \cdot \text{real} (V_{\text{out}} \cdot I_{\text{load}})) + 30 \quad (2.1)$$

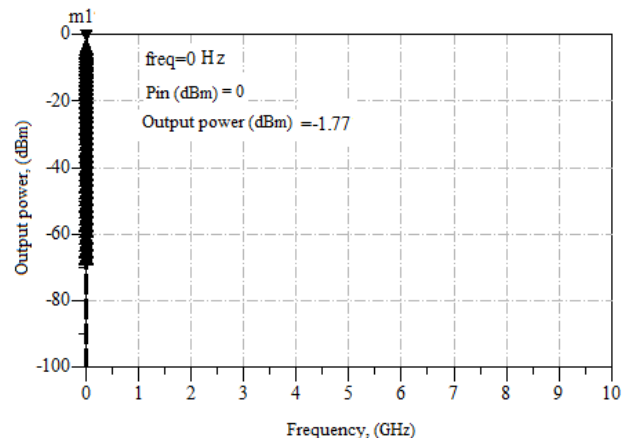


Fig. 4: DC output power (dBm) at 0 GHz.

Figure 5 shows the circuit's efficiency in Fig. 2. As can be seen, four generators (Vin1, Vin2, Vin3, and Vin4) are connected, expressing multi-input and single Vout of output (MISO). Equation (2.2) was used to calculate the simulation of the efficiency. It achieved 10.4%, 39.7%, 63.9%, and 77% for input power values of -25, -15, -5, and 0 dBm, respectively.

The efficiency is ( $\eta$ ) =

$$\frac{(P_{dc\_Watt})}{Input\ power1 + Input\ power2 + Input\ power3 + Input\ power4} \cdot 100\% \quad (2.2)$$

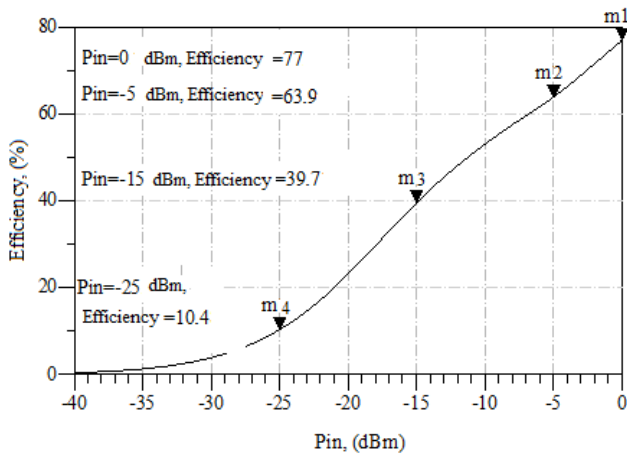


Fig. 5: Simulation efficiency of system.

Figure 6 shows values of -20, -10, and 0 dBm for input power, giving outputs of 0.14, 0.76, and 2.57 V, respectively. This result demonstrates the rectenna behaviour in terms of how the increase in input power affects output voltage.

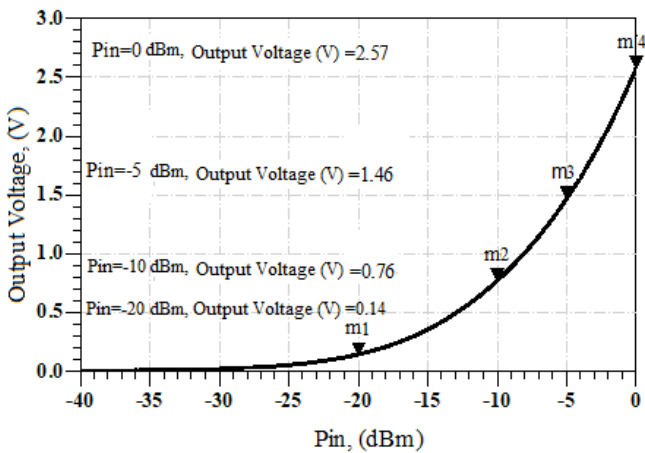


Fig. 6: Output voltage (V) versus input power (dBm).

### 3. Fabrication and Measurement

This experiment involved the conversion of RF to DC from ambient RF sources. It used a T-junction power combiner coupler integrated with a parallel voltage doubler. Figure 7 shows a photo of the fabricated prototype with dimensions (5 × 3 × 1) cm. It acts as a dual-voltage doubler for an RF-DC conversion circuit.

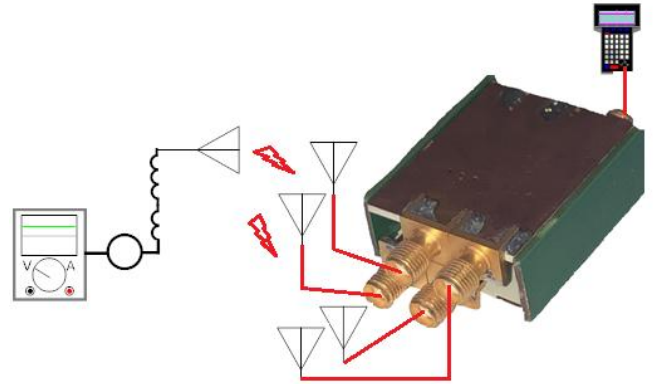


Fig. 7: Photo of fabricated circuit.

The measured setup indoors is depicted to show the path loss between the one transmission antenna (Tx) and the proposed circuit. It consists of four antenna receivers, as illustrated by far-field Equations (3.1) and (3.2), with the Friis equation in Equation (3.3). The measurement results for a power receiver are presented in Table 4.

For a far field of 2 GHz:

$$d \geq \frac{2D^2}{\lambda} = \frac{2 \cdot 12^2}{15} = 19.2 \text{ cm} \quad (3.1)$$

For a far field of 6 GHz:

$$d \geq \frac{2D^2}{\lambda} = \frac{2 \cdot 12^2}{5} = 57.6 \text{ cm} \quad (3.2)$$

$$P_{rx} = P_{tx} + G_{tx} + G_{rx} + 20 \cdot \log_{10} \left( \frac{\lambda}{4d} \right) \quad (3.3)$$

In a case where  $\lambda = 0.15 \text{ m}$ ,

$$= -5 + 6 + 15 + (-24.12) = -8.12 \text{ dBm.}$$

In a case where  $\lambda = 0.05 \text{ m}$ ,

$$= -5 + 6 + 15 + (-43.21) = -27.21 \text{ dBm.}$$

Table 4. Measurements of Output Power

Input power $P_{in} = -5$ dBm and distance $d = 1$ m			
f (MHz)	Output power ( $\mu$ W)	f (MHz)	Output power ( $\mu$ W)
1650	10	6550	10
1700	8	6600	1
1750	3	6650	6
1800	6	6700	10
1850	6	6750	3
1900	10	6800	3
1950	10	6850	5
2000	32	6900	5
2050	10	6950	8
2100	3	7000	13

The cellular base transceivers in [1], for both urban and suburban areas, presented effective radiation power (ERP), transmitting around 50 dBm for wireless electronic devices, which can cover a longer distance. In the lab, indoor radiation power of  $P_{tx} = -5$  dBm is used. In the shorter range of 1 m, the proposed system can be employed when  $G_{tx} = 6$  dB and  $G_{rx} = 9$  dB. The system is able to operate from 0.5 to 2.5 V, with an initial input sensitivity of  $-13$  dBm. The proposed circuit depends on the voltage consumer required of the device application. The Friis equation in (3.3) is applied to find the proposed receiver circuit's output power. In this case, we take  $\lambda = 0.15$  m and use a distance  $d$  of 500 m:

$$P_{rx} = 50 + 15 + 6 + (-92.44) = -21.44 \text{ dBm.}$$

The distance  $d$  is 250 m when  $\lambda = 0.15$  m.

$$P_{rx} = 50 + 15 + 6 + (-86.42) = -15.42 \text{ dBm.}$$

The limitations of a rectifier are required to perform the output DC voltage for WPT [6]. These limitations and assumptions depend on the output voltage, as seen in Fig. 6. When the target is 1.5 V, it needs input power  $-7$  dBm, which can be applied to powering sensors. Therefore, the circuit can be very useful for WPT. For power harvesting, the distance must be calculated by the Friis equation for power receiver.

Table 5 shows a comparison between this work and some of the previous research mentioned in the Introduction section. The targeted parameters are related to wireless power transfer in ISM applications [12-13]. A broad-band rectenna  $1 \times 4$  quasi-Yagi antenna array is formed with a T-junction power divider for GSM-1800 and UMTS-2100 bands in [14].

It obtained 40% PCE, 224 mV output DC voltage at 5 k $\Omega$  resistor, with input power of  $-3.4$  dBm. In [15], a dual-band rectenna array was fabricated using a power combiner in two pairs for powering an IoT application sensor. It operates at frequencies of 895-925 MHz and 1.6-2.65 GHz. The resistor load is 2200  $\Omega$  and input power is  $-10$  dBm. The efficiency obtained is 55% for the GSM900, 41.5% for the DCS1800, and 33% for the WiFi bands. Energy harvesting can be used to power sensor nodes in many applications, such as healthcare monitoring, environment monitoring, and structural monitoring [16].

Table 5. Comparison Between This Work and Other Reported Work

Ref.	Freq (MHz)	Input power $P_{in}$ (dBm)	Efficiency $\eta$ (%)	Rload ( $\Omega$ )	Size (mm)
[4]	915 2450	-9	37 30	2200	60 $\times$ 60
[5]	915 2450	-13.5 -11	40.7 56.2	2200	61.5 $\times$ 44.46
[6]	2110 to 2170	17	57%	300	70 $\times$ 50
This work	1650 to 2100 & 6550 To 7000	0 -15	77 39.7	5000	50 $\times$ 30 $\times$ 10

#### 4. Conclusion

This work presented an enhanced structural implementation that operates using two opposing layers and with fewer components. It combines double dual input RF in multi-dimensions and accumulates through the rectifier output. The rectifier was verified to cover most ISM frequency bands, especially for WPT. The PCE achieved 77% efficiency at 0 dBm input power level with low loss of DC output power, which is essential for self-powering of wireless electronic devices.

As the circuit facility demonstrates, our design for the output voltage obtained 2.57 V of RF to DC for WPT, giving it confirmable uses in telecommunication systems such as RFID and IoT sensor applications. Further, the design's

effectiveness makes it a prospective candidate for facilitating self-sustainability in communication systems, mainly as part of the green communication business.

Overall, in comparison with the previous research endeavours mentioned earlier in the paper, this work has improved several parameters such as efficiency, broadband, and robustness. Moreover, the circuit is compact in a box, offering protection to the components from environmental impacts, and is low-cost, lightweight, and small in size.

Future work will focus on improving multi-rectifier combinations to increase output power.

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