

Underwater Acoustic Noise Cancellation Scheme Combining Kalman Filter With Adaptive FxNLMS

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Abstract: Underwater environments are more challenging than that of terrestrial. The performance of a controller or the augmented system as a whole depends on the real measured data, so noise on data readings can be fatal. To effectively and adaptively control lower and higher frequency noise, the Active Noise Cancellation (ANC) was developed. Designing a system and the parameter of modified FxLMS for reducing noise, and disturbance of sensor data is the primary focus of this paper. The required equation will be analyzed and discussed briefly. Moreover, the system will be simulated in MATLAB and then the filtered result will be analyzed. Based on the simulation results, the proposed model can filter out signals with noise, particularly when there is a significant variation in the data and no knowledge of the noise frequency that might affect sensor readings.

Keywords: FxLMS; Kalman Filter; Signal Denoising; Signal Enhancement

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1. Introduction

It becomes very challenging to eliminate noise without losing some signal information if noise and signal share the same frequency band. As a result, it is difficult to remove noise from a signal in an underwater environment without losing some of its original characteristics. Acoustic noise cancellation (ANC) has received a lot of attention as a technique for removing noise from signals. The adaptive filter is a critical component of ANC because it provides noise reduction without prior knowledge of the noise and signal [1]. In ANC, a 180-degree phase signal (anti-noise) is generated and used to interfere destructively with the unnecessary noise. Bernard Widrow et al. pioneered the core concept [2].

Various methods have been proposed in order to improve the performance of ANC. RLC, LMS, and their variants (NLMS, VLMS, and so on) are popular because they have fewer complications [1]. Despite maintaining an excellent rate of convergence, the RLS algorithm fails to track the Estimation because the algorithm is dependent on its model, input data as the computation progresses, and the correlation matrix [3]. As a result, the LMS and its successor algorithms are most likely the most widely used algorithms. The Filtered-X LMS (FxLMS) algorithm is a simple variant of the LMS algorithm, which was developed independently in the context of adaptive control systems, and originally introduced as a modification in applications where an intervening system exists in the error path [4] [5].

LMS is based on the steepest descent method but does not account for secondary path effects, making it impossible to generate a precise anti-noise signal. The FxLMS algorithm is computationally simple and includes secondary path effects [6]. Several ANC algorithms with improved convergence properties have been proposed, including ANC systems in the

frequency domain (Kuo & Taherzadeh, 1997); Recursive Least Squares (RLS) based algorithms called filtered-x RLS (FxRLS) (Kuo & Morgan, 1996) and Filtered-x fast transversal filter (FxFTF) (Bouchard & Quednau, 2000); Lattice ANC systems (Park & Sommerfeldt, 1996) and Infinite impulse response (IIR) filter based LMS algorithms called filtered-u recursive LMS (FuRLMS) (L. J. Eriksson and Allie, 1987), and filtered-v algorithms (Crawford & Stewart, 1997). The basic problems in the above approaches are inherent stability problems in IIR-based structures, increment in the computational requirement and numerical instability problems in RLS based ANC systems [7]. For these reasons, FxLMS remains a viable option for ANC applications.

The step size is the most inherent feature of the Least Mean Squares (LMS) algorithm that FxLMS inherited, and it requires careful adjustment. The Small step size, required for small excess mean square error, results in slow convergence. Large step size, needed for fast adaptation, may result in loss of stability. For controlling the step size and making it variable rather than fixed, we used the Kalman filter. Kalman filter was proposed by R. E. Kalman in 1960 [8] is popular for having easy computation, memory requirements and good capability on overcoming noises. There are various types of Kalman Filter, such as standard Kalman Filter, Extended Kalman Filter, Unscented Kalman Filter etc [9]. The paper used standard Kalman filter since it contains enough part of equation for noise reducing.

The paper is organized in the following way: Section II presented all the necessary equations of FxLMS and linear Kalman filter, and also devoted to the developing of a modified FxLMS with Kalman filter to reduce active noises. Section III is devoted to the simulation and discussions of the obtained results. The conclusion section closes the paper.

2. Materials and Methods

The aim of this paper is to demonstrate a simulation, where we proposed a noise reduction model by modifying existing FxLMS algorithm, employing Kalman filter. The step size of FxLMS needs to be carefully adjusted. We replaced step size μ with kalman gain in each iteration. There are two step sizes in FxLMS, one in LMS part and another in the secondary noise path. So, in that case, the whole simulation based on three assumptions: using μ during LMS and Kalman-Gain at secondary noise path, using Kalman-Gain during LMS and μ at secondary noise path, and using Kalman-Gain during both in LMS and in secondary noise path. Finally, we compared data with novel FxLMS algorithm based on Signal to noise ratio (SNR). Our experimental data shows that after using kalman, it filtered noisy signal more efficiently than

Widrow, Shur & Shaffer proposed the integration of a secondary path model in the reference signal path (from speaker to error microphone) [11]. Figure 1 shows the block diagram of FXLMS algorithm on how the noise reduction algorithm works and the definition of each symbol is shown in Table 1.

ANC has commonly used in two different configurations of the FxLMS algorithm, one is a

before. All the simulations are carried away by MATLAB R2015b on a Windows 10 PC (x64) with an Intel i3-7100U CPU and an Nvidia GeForce 920MX GPU card.

2.1. FxLMS Algorithm

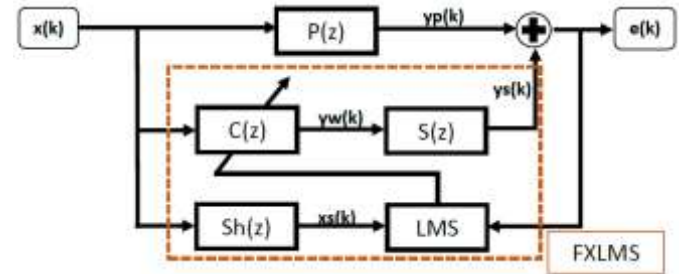


Figure 1. Block diagram of FxLMS algorithm

feedforward ANC approach proposed by Olson and May in 1953 [12], in this model, a microphone is used as an error sensor, and also as a reference sensor. The second is a feed-forward ANC approach, which uses two sensors, an error sensor, and a reference sensor. This setup is used for narrow-band noise control using a non-acoustic reference sensor [13].

Table 1. Symbols and Definitions for Figure 1

Symbols	Definitions
$x(k)$	Noise signal
$xs(k)$	Noise signal combined with assumed $Sh(z)$ based on $S(z)$
$P(z)$	Primary path transfer function
$yp(k)$	Primary noise signal at the error microphone
$e(k)$	Modified error signal
$S(z)$	Secondary path transfer function
$C(z), Sh(z)$	Controller for the FxLMS algorithm
$yw(k)$	Generated noise based on $C(z)$ controller
$ys(k)$	Output of adaptive filter

Figure 1 shows the general how the feedforward approach uses a different microphone to measure the signal at the output. It is significant to know the software elements that are part of the ANC controller, these are the LMS adaptive algorithm that updates the coefficients of the $W(z)$ adaptive filter, which in this case is represented as an FIR filter. The $C(z)$ filter represented the secondary path estimation or the transfer function between the secondary source (control source) and the error microphone. To derive the FxLMS algorithm, a similar method of the LMS algorithm is utilized but with the steepest descent, the following update equation can lead to this minimization:

$$W_{New} = W_{Old} + \mu \nabla J(n) \tag{1}$$

Where W is the controller weight error, μ is an adaption step size (scalar), $J(n)$ is the power of error signal.

The derivation of $\nabla J(n)$,

$$J(n) = E\{e^2(n)\}$$

Where $E\{\cdot\}$ denotes statistical expectation operator and $E\{\cdot\}$ is a theoretical function. To avoid this operator, $J(n)$ is approximated by

$$J(n) \approx e^2(n)$$

Then, estimate $\nabla J(n)$ as follows,

$$\begin{aligned} \nabla J(n) &= \nabla e^2(n) \\ \nabla J(n) &= 2e(n)\nabla e(n) \end{aligned}$$

2)

Now to estimate $\nabla e(n)$, the derivation is as follows based on the block diagram,

$$e(n) = d(n) + s(n) * y(n)$$

Where $s(n)$ is the secondary path impulse response.

$$\nabla e(n) = s(n) * \nabla y(n) \tag{3}$$

Now to estimate $\nabla y(n)$, the derivation is as follows based on the block diagram,

$$y(n) = W^T x(n)$$

Where W is the controller weight vector and x is the reference signal tap vector (of the same length as the controller length)

Now $\nabla y(n)$ can be expressed by,

$$\nabla y(n) = \frac{\partial y(n)}{\partial W}$$

$$\nabla y(n) = x(n) \tag{4}$$

Substitute (4) into (3)

$$\nabla e(n) = s(n) * x(n) \tag{5}$$

Substitute (5) into (2)

$$\nabla J(n) = 2e(n).s(n) * x(n) \tag{6}$$

Substitute (6) into (1)

$$W_{New} = W_{Old} + 2\mu e(n).s(n) * x(n) \tag{7}$$

The reference signal is filtered by $\hat{s}(n)$ before passing through the standard LMS algorithm. Therefore, resulting the compensation for secondary path. $\hat{s}(n)$ should be estimated through off-line or online secondary path techniques. If $\hat{s}(n)$ denotes an estimate of $s(n)$, then

$$W_{New} = W_{Old} + 2\mu e(n).\hat{s}(n) * x(n)$$

OR

$$W_{New} = W_{Old} + 2\mu e(n).x_f(n)$$

The stability of the FxLMS algorithm is highly dependent on the $x_f(n)$ power where it directly proportional to the step-size μ . So, Step-size is indirectly proportional to the steady state performance.

FxLMS is simple, fast, and surprisingly robust. Despite its straightforwardness, FxLMS acquired the most central feature of the Least Mean Squares (LMS) algorithm is the step size, and it undoubtedly requires precise adjustment. To properly control step size, we utilized the Kalman filter.

2.2. Kalman Filter

The paper will use a standard Kalman filter since it contains enough parts of the equation for noise cutting. Kalman Filter has two parts, the predicted part, and the update part. The standard Kalman Filter equation is shown in (11) – (15).

Predict:

$$\hat{x}_{t|t-1} = F_t \hat{x}_{t-1|t-1} + B_t u_t \tag{8}$$

$$P_{t|t-1} = F_t P_{t-1|t-1} F_t^T + Q_t \tag{9}$$

Update:

$$\hat{x}_{t|t} = \hat{x}_{t|t-1} + K_t (y_t - H_t \hat{x}_{t|t-1})$$

10)

$$K_t = P_{t|t-1} H_t^T (H_t P_{t|t-1} H_t^T + R_t)^{-1} \tag{11}$$

$$P_{t|t} = (1 - K_t H_t) P_{t|t-1} \tag{12}$$

where x is estimated state, F is state transition matrix, u is control variables, B is control matrix, P is state variance matrix, Q is process variance matrix, y is measurement variables, H is measurement matrix, K is Kalman gain, R is measurement matrix, $t|t$ is current time period, $t - 1|t - 1$ is previous time period, and $t|t - 1$ is intermediate steps.

To implement Kalman Filter algorithm, so that it can be used to reduce noise of sensor-readings, some adjustments for the conditions are needed. Those adjustments are as follows [14].

2.2.1. Predicting the state

On this stage, adjustments are done in (11) by giving the score $F_t = 1$ because there is no state transition. Thus, reducing the system's input component B_t because the used system does not have any input u_t . The adjusted equation is shown in (6).

$$x_{t|t-1} = x_{t-1|t-1} \tag{13}$$

2.2.2. Predicting the error

Since $F_t = 1$, then (9) becomes (14)

$$P_{t|t-1} = P_{t-1|t-1} + Q_t \tag{14}$$

2.2.3. Updating the state value

From (10), $H_t = 1$ since the sensor data that will be filtered is only consisted of one sensor reading. Hence, the equation can be written as (15).

$$x_{t|t} = x_{t|t-1} + K_t (y_t - x_{t|t-1}) \tag{15}$$

2.2.4. Calculating the gain of Kalman

Since $H_t = 1$, then (11) can be written as (16)

$$K_t = P_{t|t-1} (P_{t|t-1} + R)^{-1} \tag{16}$$

2.2.5. Updating the error value

Since $H_t = 1$, then (12) can be written as (17)

$$P_{t|t} = (1 - K_t) P_{t|t-1} \tag{17}$$

The Kalman Filter equation can be modified to reduce sensor reading noise once the necessary adjustments have been performed. The weights given to the data and the current-state estimate are represented by the Kalman-gain (at eq. 19), which can be "adjusted" to get a specific performance. We replaced the step-size μ in the FxLMS with this Kalman gain so that the step

size is flexible according to the signal elements rather than being fixed.

As we replaced step size, μ with Kalman-gain in the original FxLMS algorithm, we needed to declare some necessary variables to calculate Kalman-gain out of the noisy signal. During calculating Kalman-gain, we had to specify the values for Q (process noise covariance) and R (measurement noise covariance).

The value of Q and R are chosen according to the system operations. Covariance Q and R states may not be in general observable but the measurements should be related to the states [16].

Q, the process noise covariance, contributes to the overall uncertainty. When Q is large, the Kalman Filter more closely tracks large changes in the data than when Q is small. The measurement noise covariance R determines how much information is used from the measurement. When R is large, the Kalman Filter considers the measurements to be inaccurate. The three images below visualize the positional data. The red lines represent the measurement data, the green lines are the estimated states. [17]

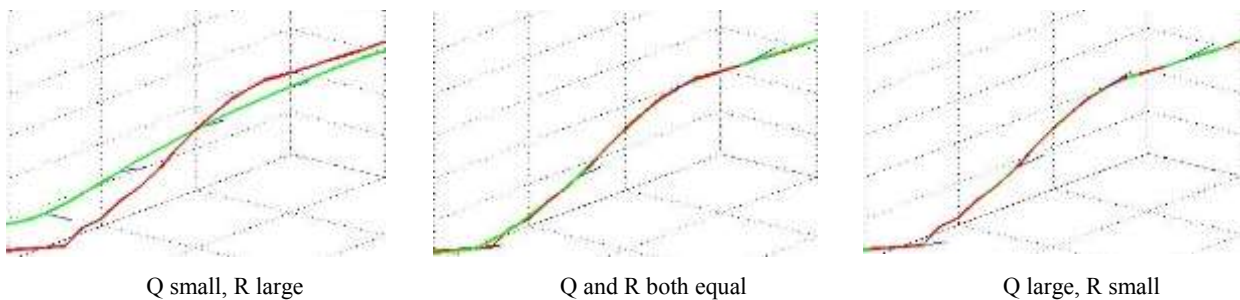


Figure 2. Relations between Q and R.

We need to balance between Q and R according to our needs. The vast majority of noise estimation methods were designed with the assumption of uncorrelated state and measurement noise in mind [18]. For example, if kalman used in tracking cars on a road, then the constant velocity model should be reasonably good, and the entries of Q should be small. Else if it is used tracking people's faces, they are not likely to move with a constant velocity, so the Q need to cranked up [19].

In [14], author used kalman filter to denoising signals. During their operations, they discovered that the greater the difference between R and Q, the greater the mean error values. Furthermore, the same R and Q values result in similar value of mean error, whatever the values of R and Q. According to their analysis, the best parameters that provide results with their original data characteristics have mean error values ranging from 40 to 55 in the table below.

Table 2. Ratio between R and Q and their yielding mean error

No. of Analysis	Kalman Filter Parameter Value		R and Q Ratio	Mean Error
	R	Q		
1	1	1	1	26.0677
2	1	0.1	10	44.7392
3	1	0.01	100	53.4466
4	10	0.1	100	53.4541
5	100	0.1	1000	56.9959

They experimentally showed that in case of signal denoising, the kalman filter yields best results if the ratio between R and Q are in 100:1. Therefore, in our operation, we kept R, Q ratio 100:1 too.

The flowchart of our modified FxLMS is given below. When we replace step size μ with kalman gain at secondary path operation, we get the best output by far.

Reference signal $x(n)$ is propagating from the source to the sensor, through the fluid medium $P(z)$. The sensor measures the arriving noise as $p(n)$. To

reduce noise, we generate another 'noise' $y(n)$ using the controller $W(z)$. We hope that it destructively interferes $x(n)$. It means that the controller has to be a model of the propagation medium $P(z)$. Least mean square algorithm is applied to adjust the controller coefficient/weight. However, there is also fluid medium $S(z)$ that stay between the actuator and sensor. We called it the secondary propagation path. So, to make the solution right, we need to compensate the adjustment process, estimate of $\hat{S}(z)$.

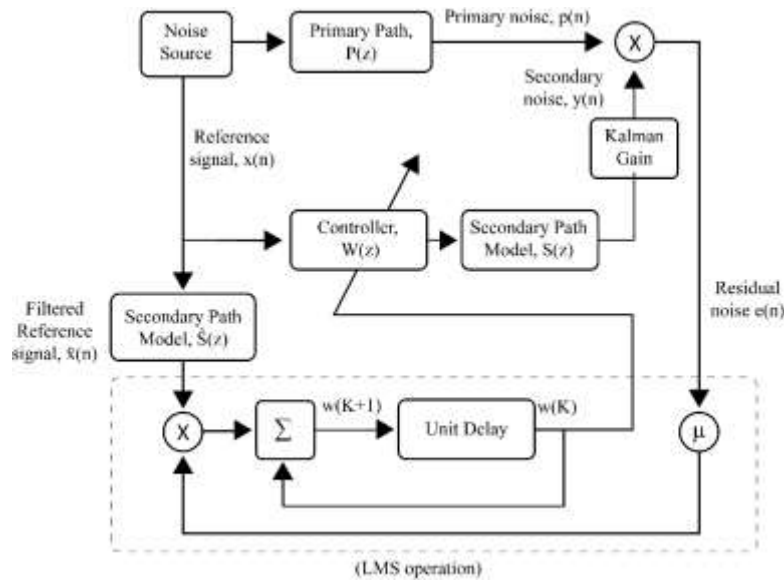


Figure 3. Modified FxLMS algorithm.

3. Results and Discussions

The research scenario is by generating a signal with noise then it will be filtered by using both FxLMS and our modified FxLMS algorithm. The step size, μ is used in two major operations in the FxLMS algorithm; one in the LMS calculating process which is run in the secondary path, and another one in the whole secondary

noise path of FxLMS. So, in that case, we run whole simulation based on three assumptions: a) Use μ during LMS and Kalman-Gain at secondary noise path, b) Use Kalman-Gain during LMS and μ at secondary noise path, c) Use Kalman-Gain during both LMS and secondary noise path.

Table 3. SNR analysis between three assumption: a) Use μ during LMS and Kalman-Gain at secondary noise path, b) Use Kalman-Gain during LMS and μ at secondary noise path, c) Use Kalman-Gain during both LMS and secondary noise path.

Use μ during LMS and Kalman-Gain at secondary noise path			Use Kalman-Gain during LMS and μ at secondary noise path			Use Kalman-Gain during both LMS and secondary noise path		
Sample no.	SNR of FxLMS	SNR of modified FxLMS	Sample no.	SNR of FxLMS	SNR of modified FxLMS	Sample no.	SNR of FxLMS	SNR of modified FxLMS
1	12.41158	13.40830	1	11.48765	11.48765	1	12.39523	13.34103
2	11.50421	12.46322	2	11.06984	11.06986	2	11.62135	12.43485
3	12.10779	12.72214	3	10.97759	10.97759	3	12.02066	12.42624
4	11.52281	12.35170	4	11.56018	11.56018	4	11.15641	11.55346
5	12.16835	12.79602	5	11.73240	11.73242	5	10.33399	9.478464
6	12.23658	13.26459	6	11.81109	11.81109	6	10.11580	11.03627
7	11.09591	11.38486	7	11.73250	11.73251	7	11.80953	12.60883
8	11.64333	11.96308	8	11.64378	11.64377	8	12.36314	13.29479
9	11.64407	12.45536	9	11.31957	11.31932	9	11.59755	12.55345
10	11.10289	11.52138	10	11.52379	11.52379	10	11.369437	12.10035

In section (a), where we used μ during LMS and Kalman-Gain at secondary noise path, yields better SNR value. In section (b), where we used Kalman-Gain during LMS and μ at secondary noise path, we almost got same SNR in novel FxLMS and in our modified FxLMS algorithm. In section (c), we used Kalman-Gain during both LMS and secondary noise path. Here we got slightly better output than that of in section (b), but in some samples, we got less SNR than novel FxLMS. So, the operation here is considered unstable.

So, we concluded, if we don't change μ during LMS operation and use Kalman-gain as a replacement of μ during secondary noise path of FxLMS, we will get more stable and better output. In figure 4, existing FxLMS and modified FxLMS plots are shown. In this particular sample we have shown, having SNR of FxLMS is 11.706468 and modified FxLMS is 12.226385.

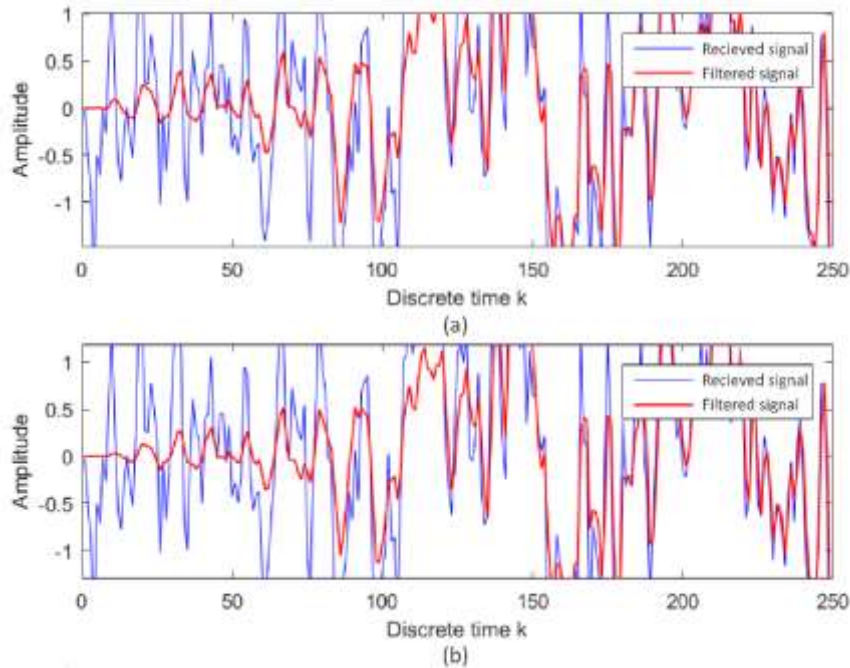


Figure 4. Plot analysis between received noisy signal and filtered signal of (a) FxLMS (SNR = 11.706468), (b) modified FxLMS (SNR = 12.226385).

As having higher SNR means more information, at figure 4(b) we got best output. In 4(a), filtered signal lost more information than that of 4(b). If we analysis noise residue plots at figure 5, modified FxLMS got

relatively less noise after each iteration, and at the end of discrete time T, noise figures are way smaller than before that indicates, our modified FxLMS capable of reducing noises much efficiently.

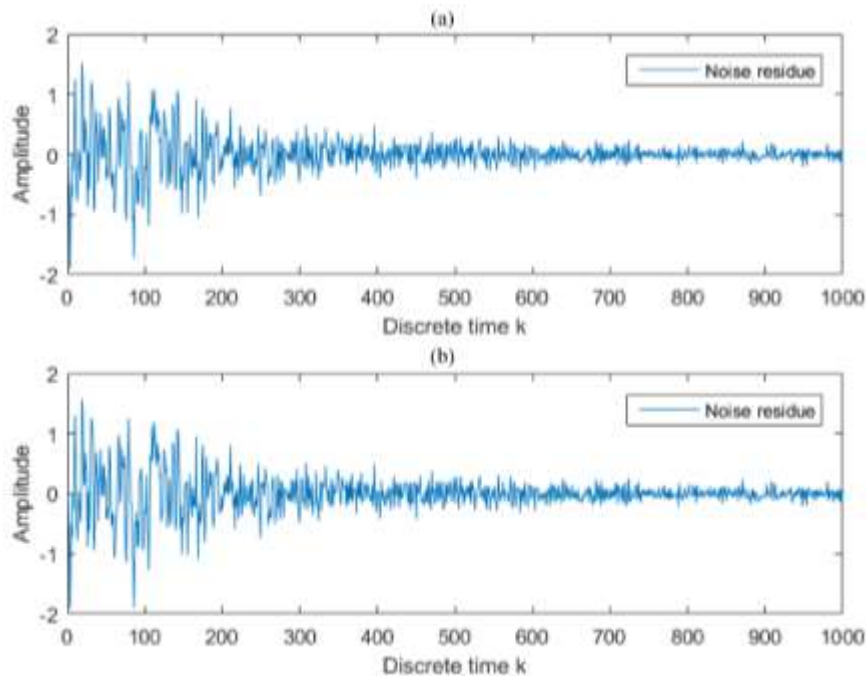


Figure 5. Plot analysis of noise residue after each filtering iteration of (a) FxLMS (SNR = 11.706468), (b) modified FxLMS (SNR = 12.226385).

4. Conclusion

To conclude, this paper demonstrated an underwater acoustic active noise cancellation (ANC)

system using Filtered-x LMS and kalman filter. Based on the simulation and test results, proposed model of modified Filtered-x LMS is able to reduce noise in received signals. Performance of modified FxLMS is

best when we keep using step size μ during LMS and replace it with kalman gain at secondary noise path calculation. The future recommendations that can be taken into consideration is by using various of kalman filter, multiple frequency tone testing, and also implementation towards the industrial system.

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